



Description

The XR76116 and XR76120 are synchronous step-down regulators combining the controller, drivers, bootstrap diode and MOSFETs in a single package for point-of-load supplies. The XR76116 has a load current rating of 15A and the XR76120 has a load current rating of 20A. A wide 5V to 22V input voltage range allows for single supply operation from industry standard 5V, 12V and 19.6V rails.

With a proprietary emulated current mode constant on-time (COT) control scheme, the XR76116/20 provide extremely fast line and load transient response using ceramic output capacitors. They require no loop compensation, simplifying circuit implementation and reducing overall component count. The control loop also provides 0.1% load and 0.1% line regulation and maintains constant operating frequency. A selectable power saving mode, allows the user to operate in discontinuous mode (DCM) at light current loads thereby significantly increasing the converter efficiency.

A host of protection features, including overcurrent, over temperature, overvoltage, short-circuit, open feedback detect and UVLO, helps achieve safe operation under abnormal operating conditions.

The XR76116/20 are available in a RoHS-compliant, green/halogen-free space-saving 5mm x 6mm QFN package.

FEATURES

- 15A and 20A step-down regulators
 - 4.5V to 5.5V low V_{IN} operation
 - 5V to 22V wide single input voltage
 - 3V to 22V operation with external 5V bias
 - $\geq 0.6V$ adjustable output voltage
- Proprietary constant on-time control
 - No loop compensation required
 - Ceramic output capacitor stable operation
 - Programmable 70ns-1 μ s on-time
 - Constant 200kHz-1MHz frequency
 - Selectable CCM or CCM/DCM operation
- Power-good flag with low impedance when power removed
- Precision enable
- Programmable soft-start
- 5mm x 6mm QFN package

APPLICATIONS

- Servers
- Distributed power architecture
- Point-of-load converters
- FPGA, DSP and processor supplies
- Base stations, switches/routers

Typical Application

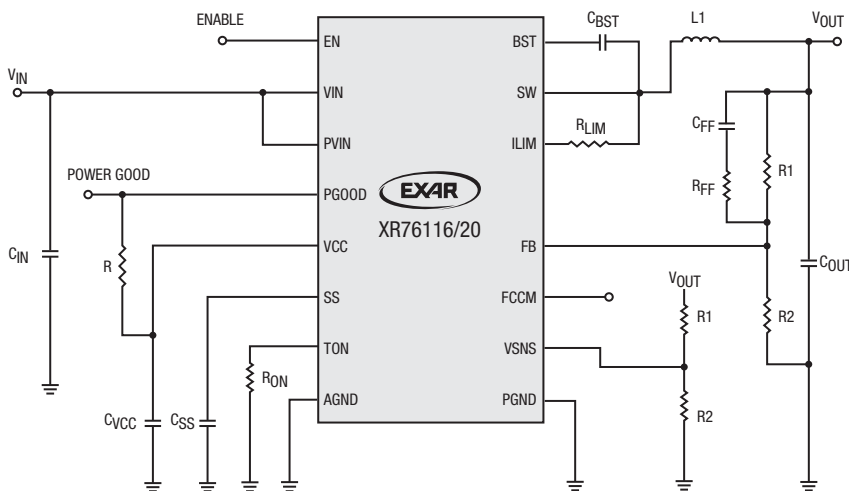


Figure 1. XR76116/20 Typical Application

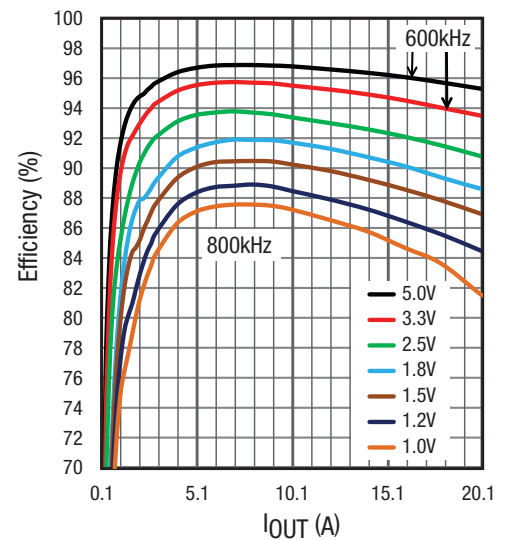


Figure 2. XR76120 Efficiency

Absolute Maximum Ratings

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. Exposure to any absolute maximum rating condition for extended periods may affect device reliability and lifetime.

| | |
|--|---------------------------------|
| PV _{IN} , V _{IN} | -0.3V to 25V |
| V _{CC} | -0.3V to 6.0V |
| BST..... | -0.3V to 31V ⁽¹⁾ |
| BST-SW..... | -0.3V to 6V |
| SW, ILIM..... | -1V to 25V ⁽¹⁾⁽²⁾ |
| All other pins..... | -0.3V to V _{CC} + 0.3V |
| Storage temperature..... | -65°C to 150°C |
| Junction temperature..... | 150°C |
| Power dissipation | Internally limited |
| Lead temperature (soldering, 10 second)..... | 300°C |
| ESD rating (HBM – human body model) | 2kV |

Operating Conditions

| | |
|--|------------------------------|
| PV _{IN} | 3V to 22V |
| V _{IN} | 4.5V to 22V |
| V _{CC} | 4.5V to 5.5V |
| SW, ILIM | -1V to 22V ⁽²⁾ |
| PGOOD, TON, SS, EN..... | -0.3V to 5.5V ⁽²⁾ |
| Switching frequency | 200kHz-1MHz ⁽³⁾ |
| Junction temperature range (T _J)..... | -40°C to 125°C |
| XR76116 package power dissipation max at 25°C | 3.8W |
| XR76120 package power dissipation max at 25°C | 4.1W |
| XR76116 package thermal resistance θ_{JA} | 26°C/W ⁽⁴⁾ |
| XR76120 package thermal resistance θ_{JA} | 24°C/W ⁽⁴⁾ |

NOTES:

1. No external voltage applied.
2. SW pin's DC range is -1V, transient is -5V for less than 50ns.
3. Recommended.
4. Measured on Exar evaluation board.

Electrical Characteristics

Specifications are for operating junction temperature of T_J = 25°C only; limits applying over the full operating junction temperature range are denoted by a •. Typical values represent the most likely parametric norm at T_J = 25°C, and are provided for reference purposes only. Unless otherwise indicated, V_{IN} = 12V, SW = AGND = PGND = 0V, CV_{CC} = 4.7uF.

| Symbol | Parameter | Conditions | • | Min | Typ | Max | Units |
|---------------------------------------|---|---|---|------|------|------|-------|
| Power Supply Characteristics | | | | | | | |
| V _{IN} | Input voltage range | V _{CC} regulating | • | 5 | 12 | 22 | V |
| | | V _{CC} tied to V _{IN} | • | 4.5 | 5.0 | 5.5 | |
| I _{VIN} | V _{IN} supply current | Not switching, V _{IN} = 12V, V _{FB} = 0.7V | • | | 0.8 | 1.3 | mA |
| I _{VCC} | V _{CC} quiescent current | Not switching, V _{CC} = V _{IN} = 5V, V _{FB} = 0.7V | • | | 0.8 | 1.3 | mA |
| I _{VIN} | V _{IN} supply current, XR76120 | f = 600kHz, R _{ON} = 49.9k, V _{FB} = 0.58V | | | 17 | | mA |
| I _{VIN} | V _{IN} supply current, XR76116 | f = 800kHz, R _{ON} = 35.7k, V _{FB} = 0.58V | | | 17 | | mA |
| I _{OFF} | Shutdown current | Enable = 0V, PV _{IN} = V _{IN} = 12V | | | 1 | | μA |
| Enable and Undervoltage Lock-Out UVLO | | | | | | | |
| V _{IH_EN} | EN pin rising threshold | | • | 1.8 | 1.9 | 2.0 | V |
| V _{EN_HYS} | EN pin hysteresis | | | | 60 | | mV |
| | V _{CC} UVLO start threshold, rising edge | | • | 4.00 | 4.25 | 4.40 | V |
| | V _{CC} UVLO hysteresis | | • | 100 | 170 | | mV |

Electrical Characteristics (Continued)

Specifications are for operating junction temperature of $T_J = 25^\circ\text{C}$ only; limits applying over the full operating junction temperature range are denoted by a •. Typical values represent the most likely parametric norm at $T_J = 25^\circ\text{C}$, and are provided for reference purposes only. Unless otherwise indicated, $V_{IN} = 12\text{V}$, $SW = AGND = PGND = 0\text{V}$, $CV_{CC} = 4.7\mu\text{F}$.

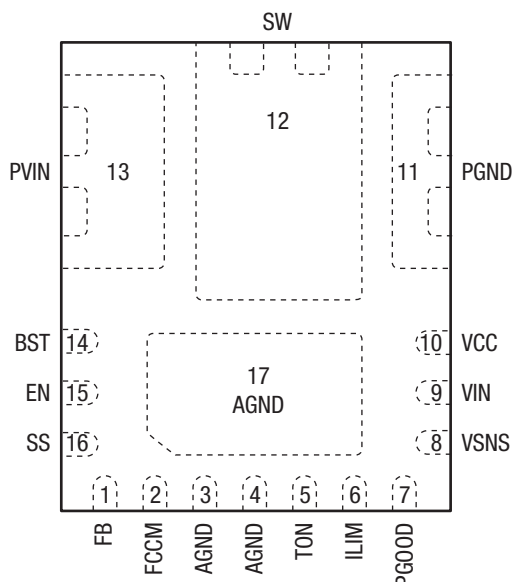
| Symbol | Parameter | Conditions | • | Min | Typ | Max | Units |
|-------------------------------|--|---|---|-------|-----------|-------|---------------|
| Reference Voltage | | | | | | | |
| V_{REF} | Reference voltage | $V_{IN} = 5\text{V} - 22\text{V}$, V_{CC} regulating | | 0.597 | 0.600 | 0.603 | V |
| | | $V_{IN} = 4.5\text{V} - 5.5\text{V}$, V_{CC} tied to V_{IN} | | 0.596 | 0.600 | 0.604 | V |
| | | $V_{IN} = 5\text{V} - 22\text{V}$, V_{CC} regulating $V_{IN} = 4.5\text{V} - 5.5\text{V}$, V_{CC} tied to V_{IN} | • | 0.594 | 0.600 | 0.606 | V |
| | DC load regulation | CCM operation, closed loop, applies to any C_{OUT} | | | ± 0.1 | | % |
| | DC line regulation | | | | ± 0.1 | | % |
| Programmable Constant On-Time | | | | | | | |
| | On-time 1 | $R_{ON} = 5.90\text{k}\Omega$, $V_{IN} = 12\text{V}$ | • | 170 | 200 | 230 | ns |
| | f corresponding to on-time 1 | $V_{OUT} = 1.0\text{V}$ | | 360 | 415 | 490 | kHz |
| | On-time 2 | $R_{ON} = 16.2\text{k}\Omega$, $V_{IN} = 12\text{V}$ | • | 425 | 500 | 575 | ns |
| | f corresponding to on-time 2 | $V_{OUT} = 3.3\text{V}$ | | 478 | 550 | 647 | kHz |
| | On-time 3 | $R_{ON} = 3.01\text{k}\Omega$, $V_{IN} = 12\text{V}$ | • | 90 | 110 | 135 | ns |
| | Minimum off-time | | • | | 250 | 350 | ns |
| Diode Emulation Mode | | | | | | | |
| | Zero crossing threshold | DC value measured during test | | | -2 | | mV |
| Soft-Start | | | | | | | |
| I_{SS_CHARGE} | Charge current | | • | -14 | -10 | -6 | μA |
| $I_{SS_DISCHARGE}$ | Discharge current | Fault present | • | 1 | 3 | | mA |
| V_{CC} Linear Regulator | | | | | | | |
| V_{CC} | Output voltage | $V_{IN} = 6\text{V}$ to 22V , $I_{LOAD} = 0$ to 30mA | • | 4.8 | 5.0 | 5.2 | V |
| | | $V_{IN} = 5\text{V}$, $R_{ON} = 16.2\text{k}\Omega$, $f_{SW} = 678\text{kHz}$ | • | 4.6 | 4.8 | | |
| Power Good Output | | | | | | | |
| | Power good threshold | | | -10 | -7.5 | -5 | % |
| | Power good hysteresis | | | | 1 | 4 | % |
| | Power good | Minimum $I_{SINK} = 1\text{mA}$ | | | | 0.2 | V |
| | Power good, unpowered | $I_{SINK} = 1\text{mA}$ | | | | 0.5 | V |
| | Power good assertion delay, FB rising | | | | 2 | | ms |
| | Power good de-assertion delay, FB falling | | | | 65 | | μs |

Electrical Characteristics (Continued)

Specifications are for operating junction temperature of $T_J = 25^\circ\text{C}$ only; limits applying over the full operating junction temperature range are denoted by a •. Typical values represent the most likely parametric norm at $T_J = 25^\circ\text{C}$, and are provided for reference purposes only. Unless otherwise indicated, $V_{IN} = 12\text{V}$, $SW = AGND = PGND = 0\text{V}$, $CV_{CC} = 4.7\mu\text{F}$.

| Symbol | Parameter | Conditions | • | Min | Typ | Max | Units |
|-------------------------------------|---|---|---|------|------|------|------------------------------|
| Mode Control (FCCM) | | | | | | | |
| | FCCM mode logic high threshold | FCCM rising | • | 2.4 | | | V |
| | FCCM mode logic low threshold | FCCM falling | • | | | 0.4 | V |
| | Input leakage current | | | | 100 | | nA |
| Open Feedback/OVP Detect (VSNS) | | | | | | | |
| | OVP trip high threshold | VSNS rising. Specified as % of V_{REF} | • | 115 | 120 | 125 | % |
| | OVP trip low threshold | VSNS falling. Specified as % of V_{REF} | • | | 115 | | % |
| | OVP comparator delay | VSNS rising | • | 0.5 | 1 | 3.5 | μs |
| | Delay to turn off power stage from an overvoltage event | VSNS rising | • | | | 3.5 | μs |
| Protection: OCP, OTP, Short-Circuit | | | | | | | |
| | Hiccup timeout | | | | 110 | | ms |
| | I_{LIM}/R_{DS} | XR76116 | | 6.30 | 7.15 | 8.00 | $\mu\text{A}/\text{m}\Omega$ |
| | I_{LIM}/R_{DS} | XR76120 | | 14.5 | 16.2 | 18.0 | $\mu\text{A}/\text{m}\Omega$ |
| | I_{LIM} current temperature coefficient | | | | 0.4 | | $\%/^\circ\text{C}$ |
| | I_{LIM} comparator offset | | | -4.7 | 0 | 4.7 | mV |
| | I_{LIM} comparator offset | | • | -8.0 | 0 | 8.0 | mV |
| | Current limit blanking | | | | 100 | | ns |
| | Thermal shutdown threshold | Rising temperature | | | 138 | | $^\circ\text{C}$ |
| | Thermal hysteresis | | | | 15 | | $^\circ\text{C}$ |
| | Feedback pin short-circuit threshold | Percent of V_{REF} , short circuit is active. After PGOOD asserts high. | • | 50 | 60 | 70 | % |
| XR76116 Output Power Stage | | | | | | | |
| | High-side MOSFET $R_{DS(ON)}$ | $I_{DS} = 2\text{A}$ | | | 7.7 | 10 | $\text{m}\Omega$ |
| | Low-side MOSFET $R_{DS(ON)}$ | $I_{DS} = 2\text{A}$ | | | 7.0 | 10 | $\text{m}\Omega$ |
| | Maximum output current | | • | 15 | | | A |
| XR76120 Output Power Stage | | | | | | | |
| | High-side MOSFET $R_{DS(ON)}$ | $I_{DS} = 2\text{A}$ | | | 7.7 | 10 | $\text{m}\Omega$ |
| | Low-side MOSFET $R_{DS(ON)}$ | $I_{DS} = 2\text{A}$ | | | 3.1 | 3.5 | $\text{m}\Omega$ |
| | Maximum output current | | • | 20 | | | A |

Pin Configuration



Pin Functions

| Pin Number | Pin Name | Type | Description |
|------------|----------|--------------------|--|
| 1 | FB | Analog | Feedback input to feedback comparator. |
| 2 | FCCM | Input | Forcing this pin logic level high forces CCM operation. |
| 3 | AGND | Analog | Signal ground for control circuitry. Connect to AGND pad with a short trace. |
| 4 | | | |
| 5 | TON | Analog | Constant on-time programming pin. Connect with a resistor to AGND. |
| 6 | ILIM | Analog | Overcurrent protection programming. Connect with a resistor to SW. |
| 7 | PGOOD | Output, Open Drain | Power-good output. Open drain to AGND. Low Z when IC unpowered. |
| 8 | VSNS | Analog | Sense pin for output OVP and open FB. |
| 9 | VIN | Analog | Supply input for the regulator's LDO. Normally connected to PV_{IN} . |
| 10 | VCC | Analog | The output of regulators LDO. It requires a $4.7\mu\text{F}$ V_{CC} bypass capacitor. For operation using a 5V rail, VCC should be tied to VIN. |
| 11 | PGND | Power | Ground of the power stage. Internally connected to source of the low-side MOSFET. |
| 12 | SW | Power | Switch node. Internally it connects source of the high-side MOSFET to drain of the low-side MOSFET. |
| 13 | PVIN | Power | Input voltage for power stage. Internally connected to drain of the high-side MOSFET. |
| 14 | BST | Analog | High-side driver supply pin. Connect a $0.1\mu\text{F}$ bootstrap capacitor between BST and SW. |
| 15 | EN | Input | Precision enable pin. Pulling this pin above 2V will enable the regulator. |
| 16 | SS | Analog | Soft-start pin. Connect an external capacitor between SS and AGND to program the soft-start rate based on the $10\mu\text{A}$ internal source current. |
| 17 | AGND PAD | Analog | Signal ground for control circuitry. |

Typical Performance Characteristics

Efficiency and Package Thermal Derating

Unless otherwise specified: $T_{AMBIENT} = 25^{\circ}C$, no airflow, $f = 800kHz$. Efficiency data includes inductor losses, schematic from the Application Information section of this datasheet.

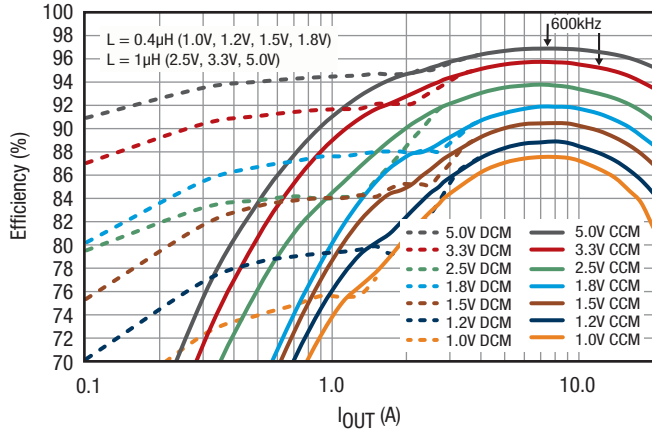


Figure 3. XR76120 Efficiency, $V_{IN} = 12V$

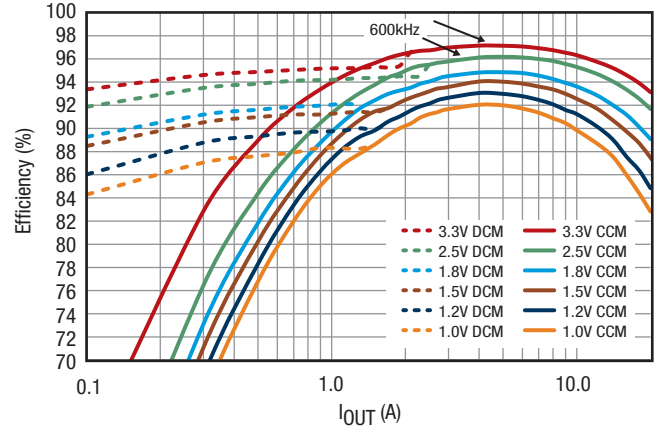


Figure 4. XR76120 Efficiency, $V_{IN} = 5V, L = 0.4\mu H$

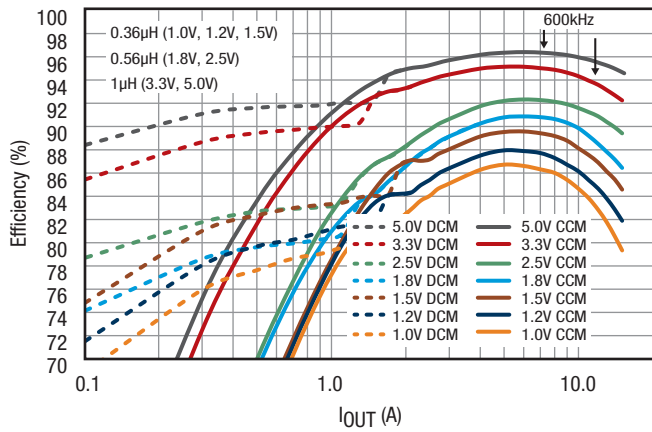


Figure 5. XR76116 Efficiency, $V_{IN} = 12V$

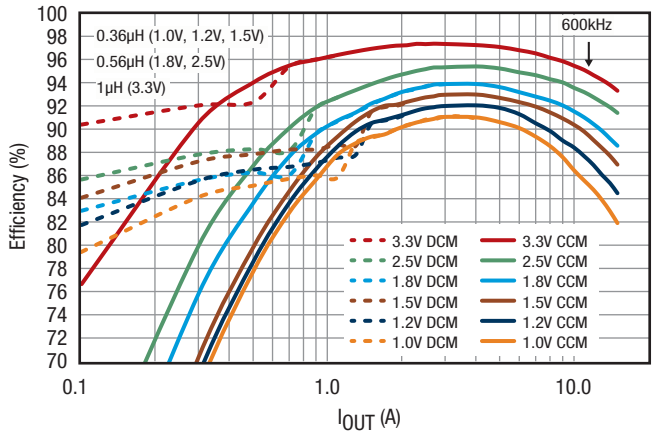


Figure 6. XR76116 Efficiency, $V_{IN} = 5V$

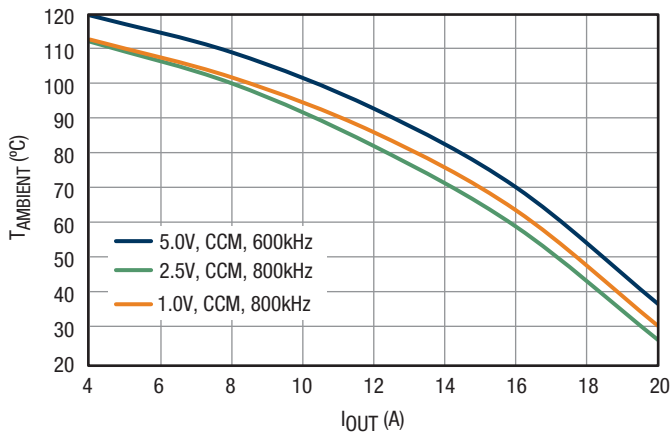


Figure 7. XR76120 Maximum $T_{AMBIENT}$ vs. I_{OUT} , $V_{IN} = 12V$, No Airflow

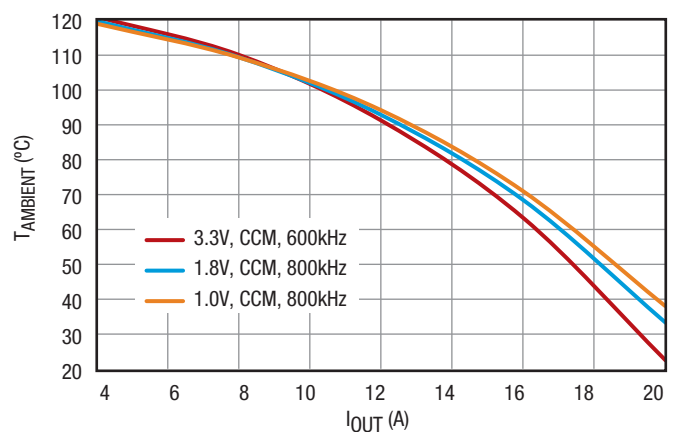


Figure 8. XR76120 Maximum $T_{AMBIENT}$ vs. I_{OUT} , $V_{IN} = 5V$, No Airflow

Typical Performance Characteristics (Continued)

Efficiency and Package Thermal Derating

Unless otherwise specified: $T_{\text{AMBIENT}} = 25^{\circ}\text{C}$, no airflow, $f = 800\text{kHz}$. Efficiency data includes inductor losses, schematic from the Application Information section of this datasheet.

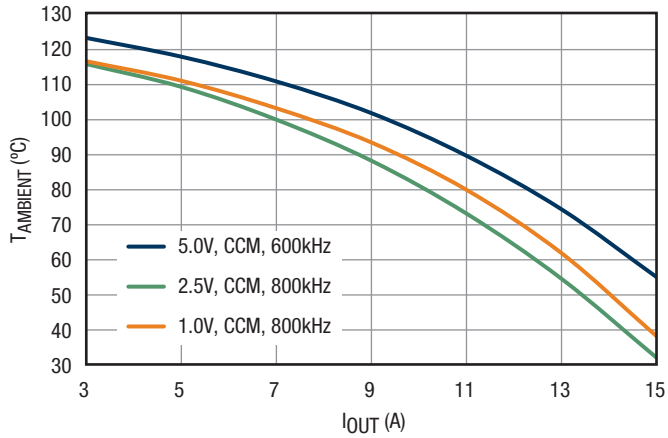


Figure 9. XR76116 Maximum T_{AMBIENT} vs. I_{OUT} , $V_{\text{IN}} = 12\text{V}$, No Airflow

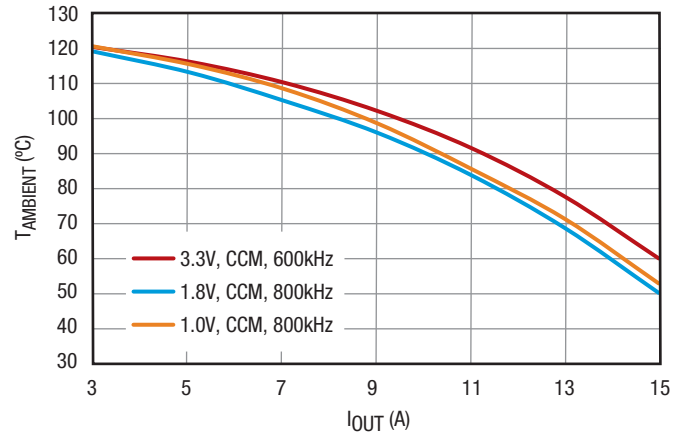


Figure 10. XR76116 Maximum T_{AMBIENT} vs. I_{OUT} , $V_{\text{IN}} = 5\text{V}$, No Airflow

Typical Performance Characteristics (Continued)

All data taken at $V_{IN} = 12V$, $V_{OUT} = 1.8V$, $f = 800kHz$, $T_A = 25^\circ C$, no airflow, forced CCM. (Unless otherwise specified). Schematic from the Applications Information section of this datasheet.

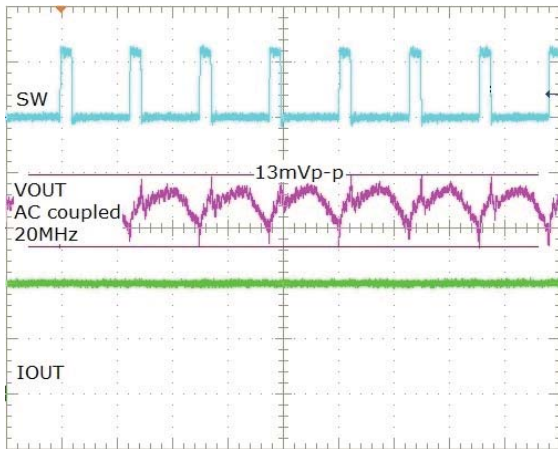


Figure 11. Steady State, $I_{OUT} = 20A$

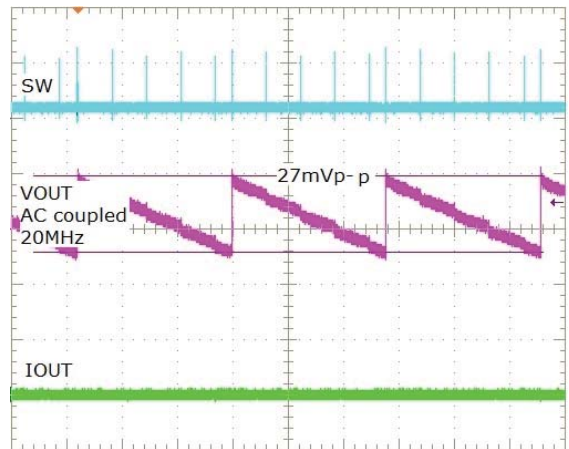


Figure 12. Steady State, DCM, $I_{OUT} = 0A$

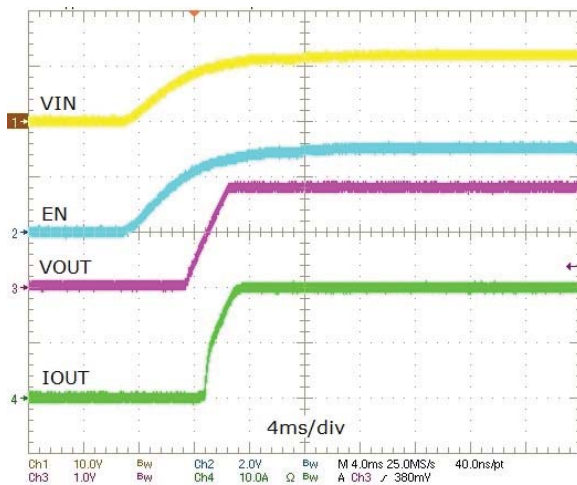


Figure 13. Power-Up, $I_{OUT} = 20A$

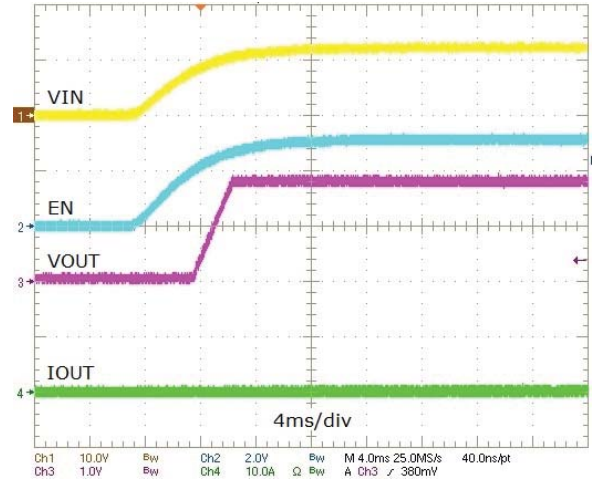


Figure 14. Power-Up, $I_{OUT} = 0A$

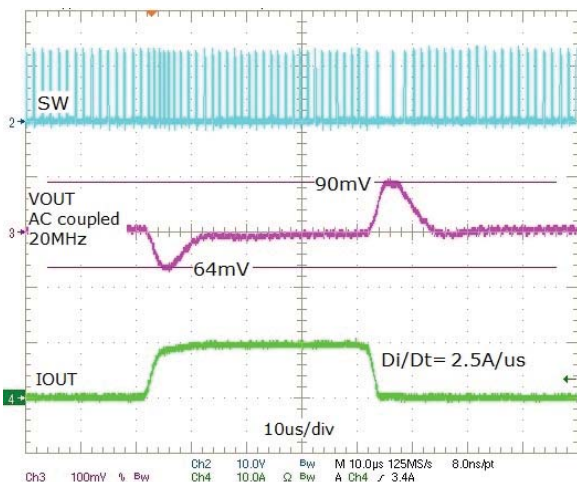


Figure 15. Load Transient, Forced CCM, 0A-10A-0A

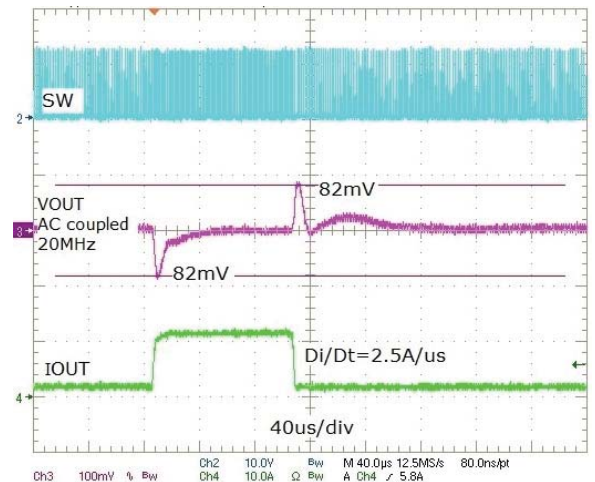


Figure 16. Load Transient, DCM, 1.8A-11.8A-1.8A

Typical Performance Characteristics (Continued)

All data taken at $V_{IN} = 12V$, $V_{OUT} = 1.8V$, $f = 800kHz$, $T_A = 25^{\circ}C$, no airflow, forced CCM. (Unless otherwise specified). Schematic from the Applications Information section of this datasheet.

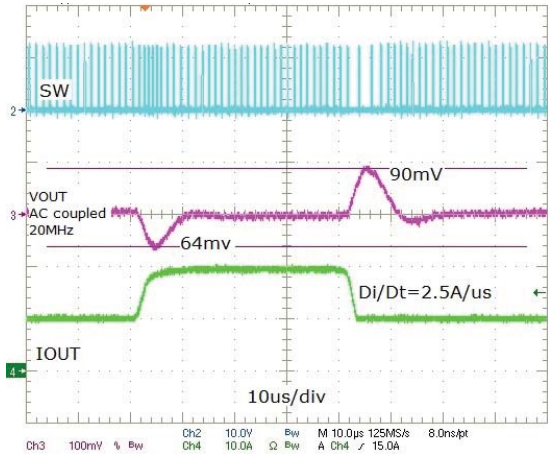


Figure 17. Load Transient, DCM or Forced CCM, 10A-20A-10A

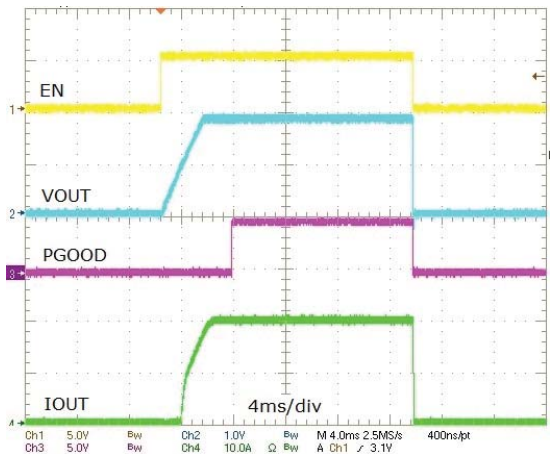


Figure 18. Enable Functionality, $V_{IN} = 12V$

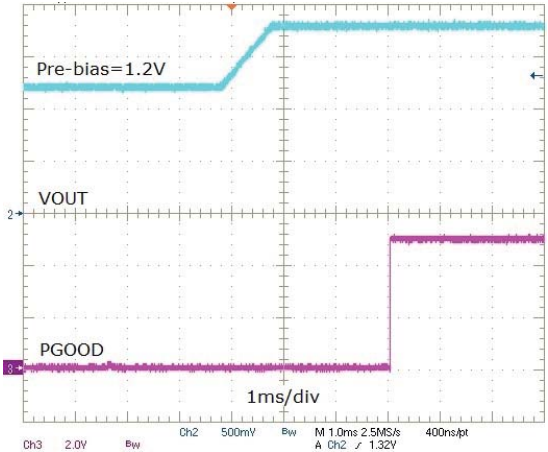


Figure 19. Power-Up with Pre-Bias Voltage, $I_{OUT} = 0A$

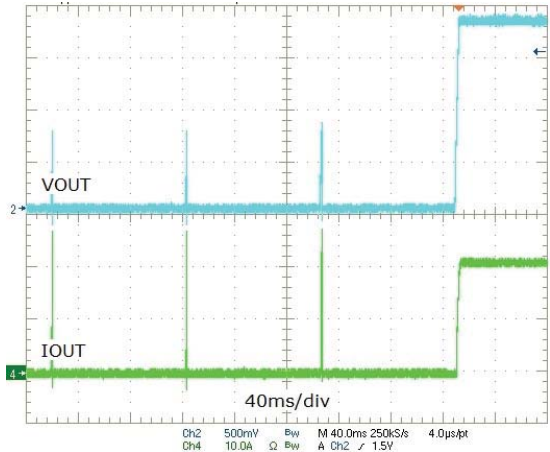


Figure 20. Short Circuit Recovery, $I_{OUT} = 20A$

Typical Performance Characteristics (Continued)

All data taken at $V_{IN} = 12V$, $V_{OUT} = 1.8V$, $f = 800kHz$, $T_A = 25^\circ C$, no airflow, forced CCM. (Unless otherwise specified). Schematic from the Applications Information section of this datasheet.

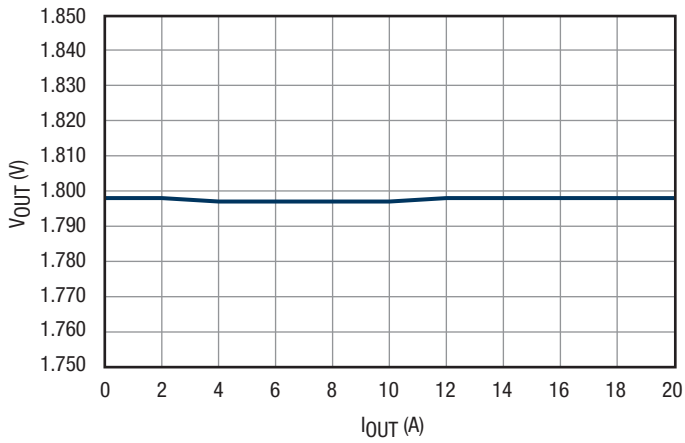


Figure 21. Load Regulation

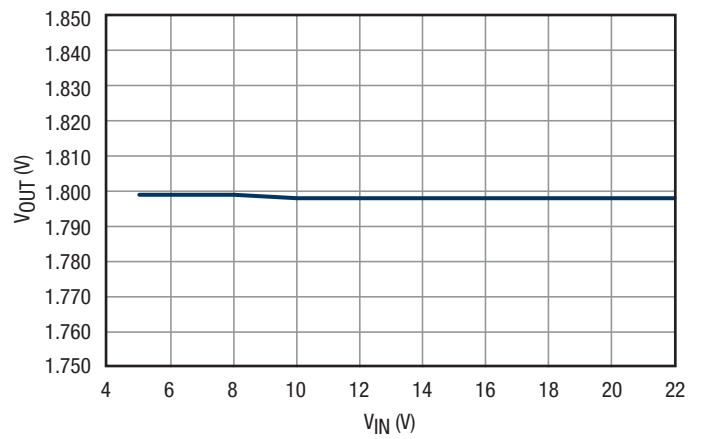


Figure 22. Line Regulation

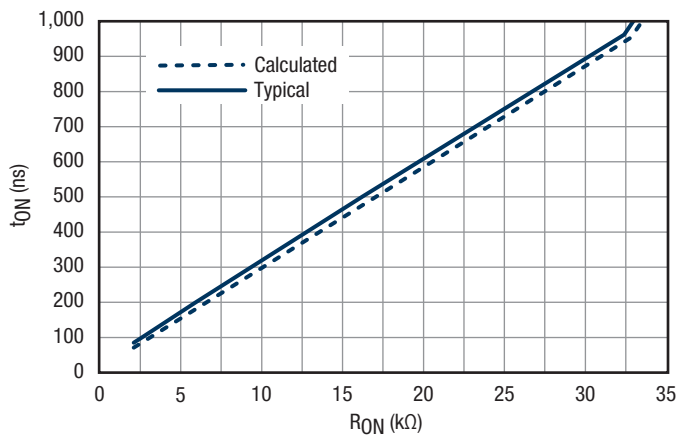


Figure 23. t_{ON} vs. R_{ON}

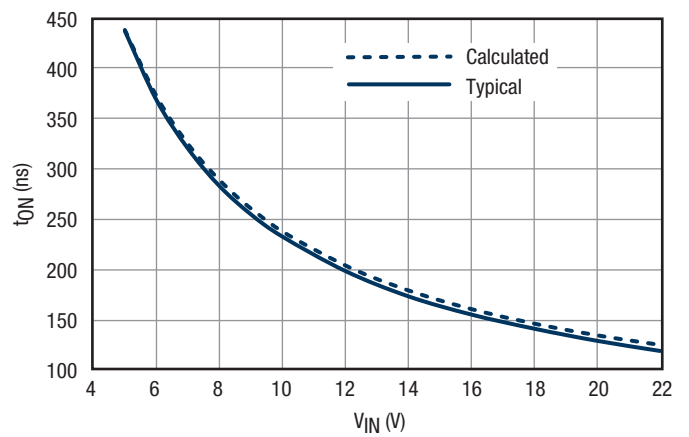


Figure 24. t_{ON} vs. V_{IN} , $R_{ON} = 5.9k\Omega$

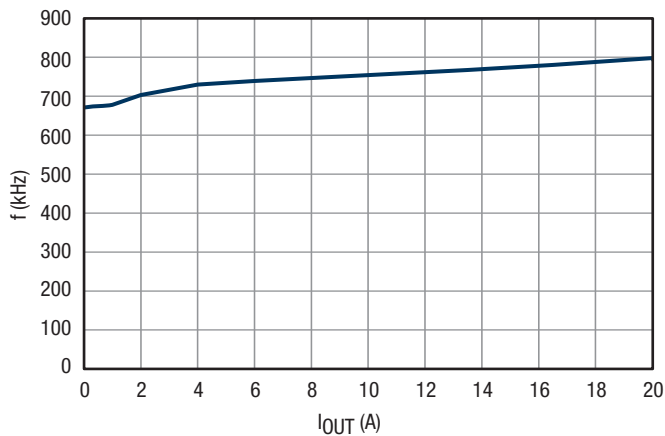


Figure 25. Frequency vs. I_{OUT}

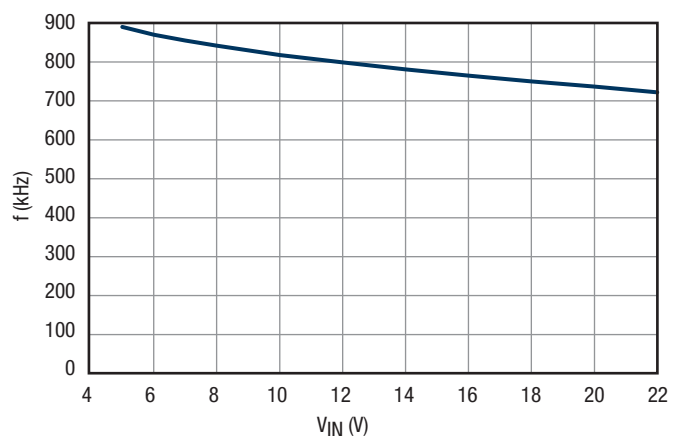


Figure 26. Frequency vs. V_{IN}

Typical Performance Characteristics (Continued)

All data taken at $V_{IN} = 12V$, $V_{OUT} = 1.8V$, $f = 800kHz$, $T_A = 25^\circ C$, no airflow, forced CCM. (Unless otherwise specified). Schematic from the Applications Information section of this datasheet.

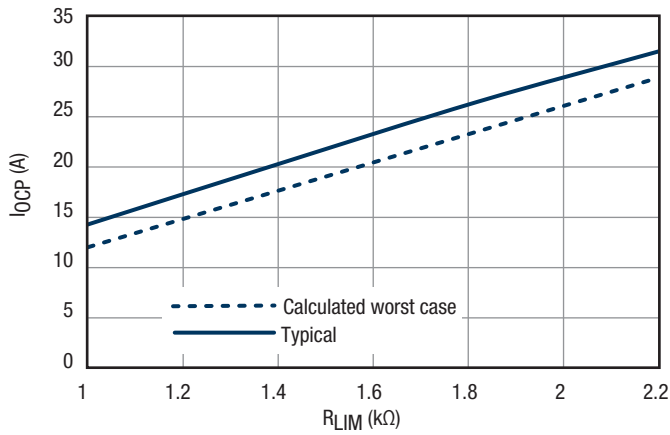


Figure 27. XR76120 I_{OCP} vs. R_{LIM}

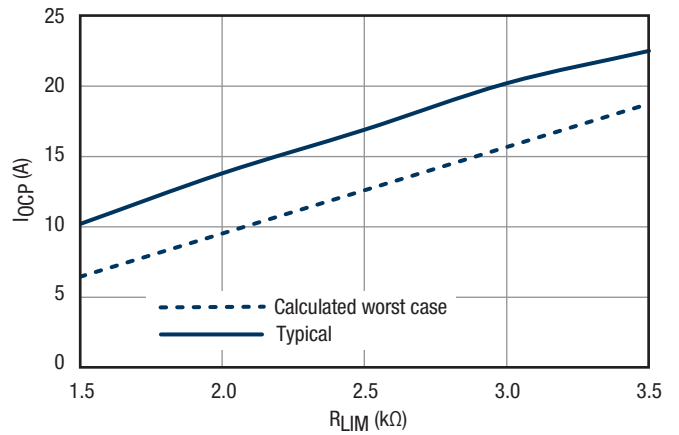


Figure 28. XR76116 I_{OCP} vs. R_{LIM}

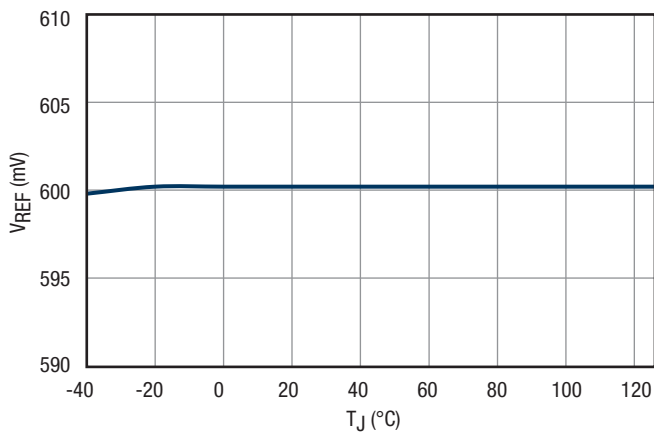


Figure 29. V_{REF} vs. Temperature

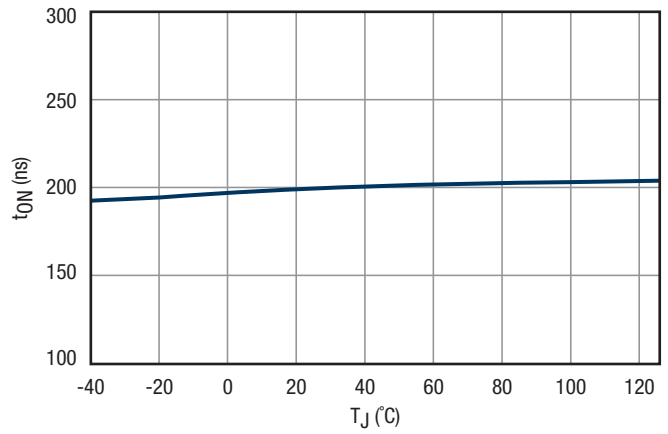


Figure 30. t_{ON} vs. Temperature, $R_{ON} = 5.9k$

Functional Block Diagram

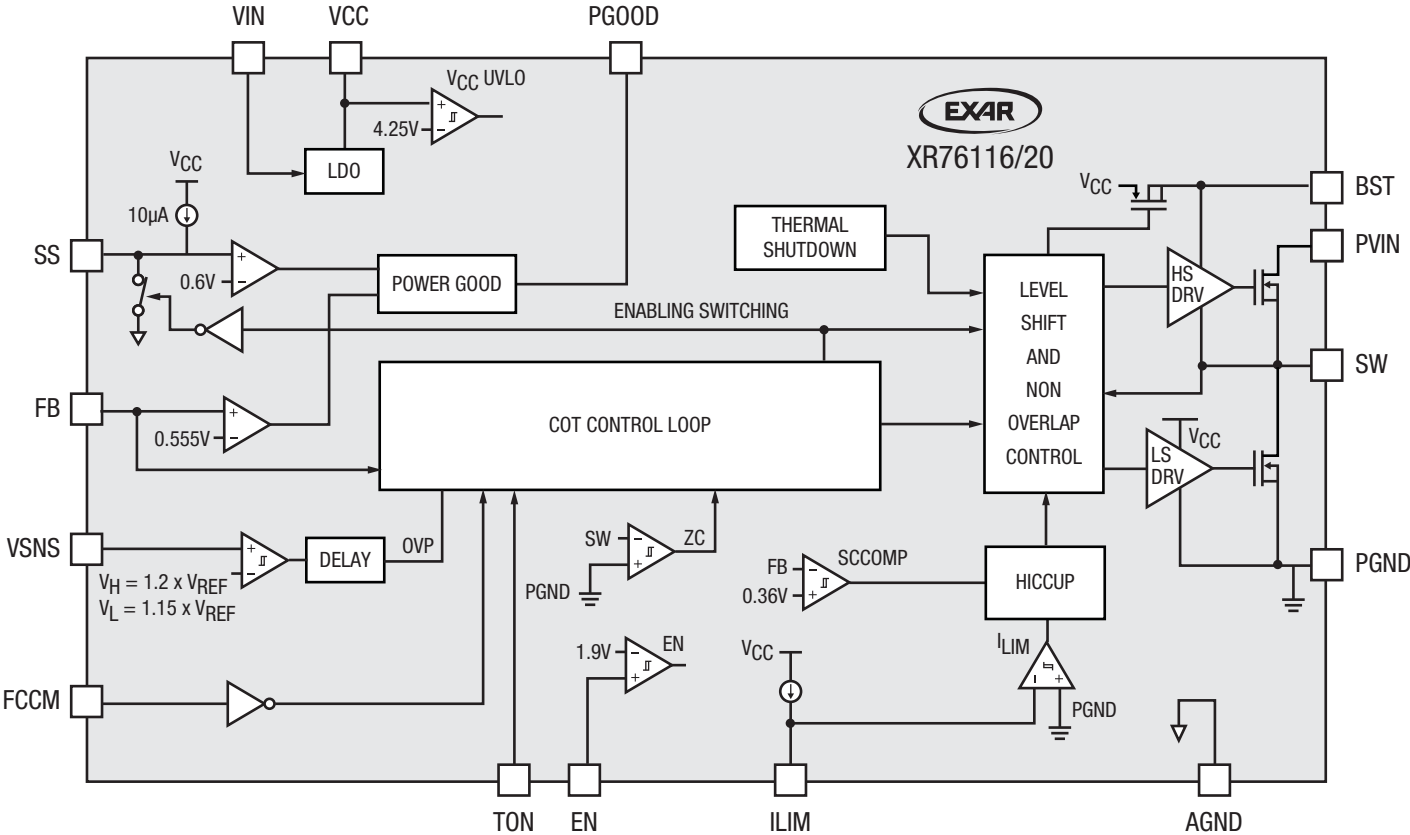


Figure 31. Functional Block Diagram

Applications Information

Detailed Operation

The XR76116/20 uses a synchronous step-down proprietary emulated current-mode Constant On-Time (COT) control scheme. The on-time, which is programmed via R_{ON} , is inversely proportional to V_{IN} and maintains a nearly constant frequency. The emulated current-mode control allows the use of ceramic output capacitors.

Each switching cycle begins with the high-side (switching) FET turning on for a preprogrammed time. At the end of the on-time, the high-side FET is turned off and the low-side (synchronous) FET is turned on for a preset minimum time (250ns nominal). This parameter is termed the minimum off-time. After the minimum off-time the voltage at the feedback pin FB is compared to an internal voltage ramp at the feedback comparator. When V_{FB} drops below the ramp voltage, the high-side FET is turned on and the cycle repeats. This voltage ramp constitutes an emulated current ramp and allows for the use of ceramic capacitors, in addition to other capacitor types, for output filtering.

Enable

The enable input provides precise control for startup. Where bus voltage is well regulated, the enable input can be derived from this voltage with a suitable resistor divider. This ensures that XR76116/20 does not turn on until bus voltage reaches the desired level. Therefore the enable feature allows implementation of undervoltage Lockout for the bus voltage PV_{IN} . Simple sequencing can be implemented by using the PGOOD signal as the enable input of a succeeding XR76116/20. Sequencing can also be achieved by using an external signal to control the enable pin.

Selecting the Forced CCM Mode

A voltage higher than 2.4V at the FCCM pin forces the XR76116/20 to operate in continuous conduction mode (CCM). Note that discontinuous conduction mode (DCM) is always on during soft-start. DCM will persist following soft-start until a sufficient load is applied to transition the regulator to CCM. Magnitude of the load required to transition to CCM is $\Delta I_L/2$, where ΔI_L is peak-to-peak inductor current ripple. Once the regulator transitions to CCM it will continue operating in CCM regardless of the load magnitude.

Selecting the DCM/CCM Mode

The DCM will always be available if a voltage less than 0.4V is applied to the FCCM pin. XR76116/20 will operate in either DCM or CCM depending on the load magnitude. At light loads DCM significantly increases efficiency as seen in Figures 3-6. A preload of 10mA is recommended for DCM operation. This helps improve voltage regulation when external load is less than 10mA and may reduce voltage ripple.

Programming the On-Time

The on-time t_{ON} is programmed via resistor R_{ON} according to following equation:

$$R_{ON} = \frac{V_{IN} \times [t_{ON} - (2.5 \times 10^{-8})]}{3.45 \times 10^{-10}}$$

A graph of t_{ON} versus R_{ON} , using the above equation, is compared to typical test data in Figure 23. The graph shows that calculated data matches typical test data within 3%.

The t_{ON} corresponding to a particular set of operating conditions can be calculated based on empirical data from:

$$t_{ON} = \frac{V_{OUT}}{V_{IN} \times 1.06 \times f \times \text{Eff.}}$$

Where:

- f is the desired switching frequency at nominal I_{OUT} .
- Eff. is the converter efficiency corresponding to nominal I_{OUT} .

Substituting for t_{ON} in the first equation we get:

$$R_{ON} = \frac{\left(\frac{V_{OUT}}{1.06 \times f \times \text{Eff.}} \right) - [(2.5 \times 10^{-8}) \times V_{IN}]}{(3.45 \times 10^{-10})}$$

Now R_{ON} can be calculated in terms of operating conditions V_{IN} , V_{OUT} , f and efficiency using the above equation.

At $V_{IN} = 12V$, $f = 800kHz$, $I_{OUT} = 20A$ and using the efficiency numbers from Figure 3 we get the following R_{ON} for XR76120:

| V_{OUT} (V) | Eff. (%) | f (kHz) | R_{ON} (k Ω) |
|---------------|----------|-----------|------------------------|
| 5.0 | 0.95 | 600 | 23.12 |
| 3.3 | 0.93 | 600 | 15.30 |
| 2.5 | 0.91 | 800 | 8.52 |
| 1.8 | 0.89 | 800 | 6.04 |
| 1.5 | 0.87 | 800 | 5.02 |
| 1.2 | 0.84 | 800 | 4.01 |
| 1.0 | 0.81 | 800 | 3.35 |

XR76120 R_{ON} for common output voltages,
 $V_{IN} = 12V$, $I_{OUT} = 20A$

Applications Information (Continued)

Overcurrent Protection (OCP)

If the load current exceeds the programmed overcurrent threshold I_{OCP} for four consecutive switching cycles, the regulator enters the hiccup mode of operation. In hiccup mode the MOSFET gates are turned off for 110ms (hiccup timeout). Following the hiccup timeout a soft-start is attempted. If OCP persists, hiccup timeout will repeat. The regulator will remain in hiccup mode until load current is reduced below the programmed I_{OCP} . In order to program overcurrent protection use the following equation:

$$R_{LIM} = \left[\frac{(I_{OCP} + (0.5 \times \Delta I_L))}{\left(\frac{I_{LIM}}{R_{DS}} \right)} \right] + 0.16k\Omega$$

Where:

- R_{LIM} is resistor value in $k\Omega$ for programming I_{OCP}
- I_{OCP} is the overcurrent value to be programmed
- ΔI_L is the peak-to-peak inductor current ripple
- I_{LIM}/R_{DS} is the minimum value of the parameter specified in the tabulated data
- $I_{LIM}/R_{DS} = 6.3\mu A/m\Omega$ (XR76116)
- $I_{LIM}/R_{DS} = 14.5\mu A/m\Omega$ (XR76120)
- $0.16k\Omega$ accounts for OCP comparator offset

The above equation is for worst-case analysis and safeguards against premature OCP. Typical value of I_{OCP} , for a given R_{LIM} , will be higher than that predicted by the above equation. Graph of calculated I_{OCP} vs. R_{LIM} is compared to typical I_{OCP} in Figures 27 and 28.

Short-Circuit Protection (SCP)

If the output voltage drops below 60% of its programmed value (i.e., FB drops below 0.36V), the regulator will enter hiccup mode. Hiccup mode will persist until short-circuit is removed. The SCP circuit becomes active at the end of soft-start. Hiccup mode and short-circuit recovery waveform is shown in Figure 20.

Over Temperature Protection (OTP)

OTP triggers at a nominal controller temperature of 138°C. The gates of the switching FET and the synchronous FET are turned off. When controller temperature cools down to 123°C, soft-start is initiated and regular operation resumes.

Overvoltage Protection (OVP)

The output OVP function detects an overvoltage condition on V_{OUT} of the regulator. OVP is achieved by comparing the voltage at VSNS pin to an OVP threshold voltage set at $1.2 \times V_{REF}$. When VSNS voltage exceeds the OVP threshold, an internal overvoltage signal asserts after 1 μ s (typical). This OVP signal latches off the high-side FET, turns on the low-side FET and also asserts PGOOD low. The low-side FET remains on to discharge the output capacitor until VSNS voltage drops below $1.15 \times V_{REF}$. Then low-side FET turns off to prevent complete discharge of V_{OUT} . The high-side and low-side FETs remain latched off until V_{IN} or EN is recycled. In order to use this feature, connect VSNS to V_{OUT} with a resistor divider as shown in the application circuit. Use the same resistor divider value that was used for programming V_{OUT} .

Programming the Output Voltage

Use a voltage divider as shown in Figure 1 to program the output voltage V_{OUT} .

$$R1 = R2 \times \left(\frac{V_{OUT}}{0.6} - 1 \right)$$

The recommended value for R2 is 2k Ω .

Programming the Soft-Start

Place a capacitor C_{SS} between the SS and AGND pins to program the soft-start. In order to program a soft-start time of t_{SS} , calculate the required capacitance C_{SS} from the following equation:

$$C_{SS} = t_{SS} \times \frac{10\mu A}{0.6V}$$

Pre-Bias Startup

XR76116/20 has the capability to startup into a pre-charged output. Typical pre-bias startup waveforms are shown in Figure 19.

Maximum Allowable Voltage Ripple at FB Pin

The steady-state voltage ripple at feedback pin FB ($V_{FB,RIPPLE}$) must not exceed 50mV in order for the Regulator to function correctly. If $V_{FB,RIPPLE}$ is larger than 50mV then C_{OUT} and/or L should be increased as necessary in order to keep the $V_{FB,RIPPLE}$ below 50mV.

Applications Information (Continued)

Feed-Forward Capacitor (C_{FF})

The feed-forward capacitor C_{FF} is used to set the necessary phase margin when using ceramic output capacitors. Calculate C_{FF} from the following equation:

$$C_{FF} = \frac{1}{2 \times \pi \times R1 \times 5 \times f_{LC}}$$

Where f_{LC} , the output filter double-pole frequency is calculated from:

$$f_{LC} = \frac{1}{2 \times \pi \times \sqrt{L \times C_{OUT}}}$$

You must use manufacturer's DC derating curves to determine the effective capacitance corresponding to V_{OUT} . A load step test (and/or a loop transient response test) should be performed and if necessary C_{FF} can be adjusted in order to get a critically damped transient load response.

In applications where output voltage ripple is less than about 3mV, such as when a large number of ceramic C_{OUT} are paralleled, it is necessary to use ripple injection from across the inductor. The circuit and corresponding calculations are explained in the Exar design note.

Feed-Forward Resistor (R_{FF})

R_{FF} is required when C_{FF} is used. R_{FF} , in conjunction with C_{FF} , functions similar to a high frequency pole and adds gain margin to the frequency response. Calculate R_{FF} from:

$$R_{FF} = \frac{1}{2 \times \pi \times f \times C_{FF}}$$

Where f is the switching frequency.

If R_{FF} is greater than $0.1 \times R1$, then instead of C_{FF}/R_{FF} , use ripple injection circuit as described in Exar design note.

Thermal Design

Proper thermal design is critical in controlling device temperatures and in achieving robust designs. There are a number of factors that affect the thermal performance. One key factor is the temperature rise of the devices in the package, which is a function of the thermal resistances of the devices inside the package and the power being dissipated.

The thermal resistance of the XR76116/20 is specified in the Operating Ratings section of this datasheet. The θ_{JA} thermal resistance specification is based on the XR76116/20 evaluation board operating without forced airflow. Since the actual board design in the final application will be different, the thermal resistances in the final design may be different from those specified.

The package thermal derating curves for the XR76120 are shown in Figures 7 and 8. These correspond to input voltage of 12V and 5V, respectively. The package thermal derating curves for the XR76116 are shown in Figures 9 and 10.

Applications Information

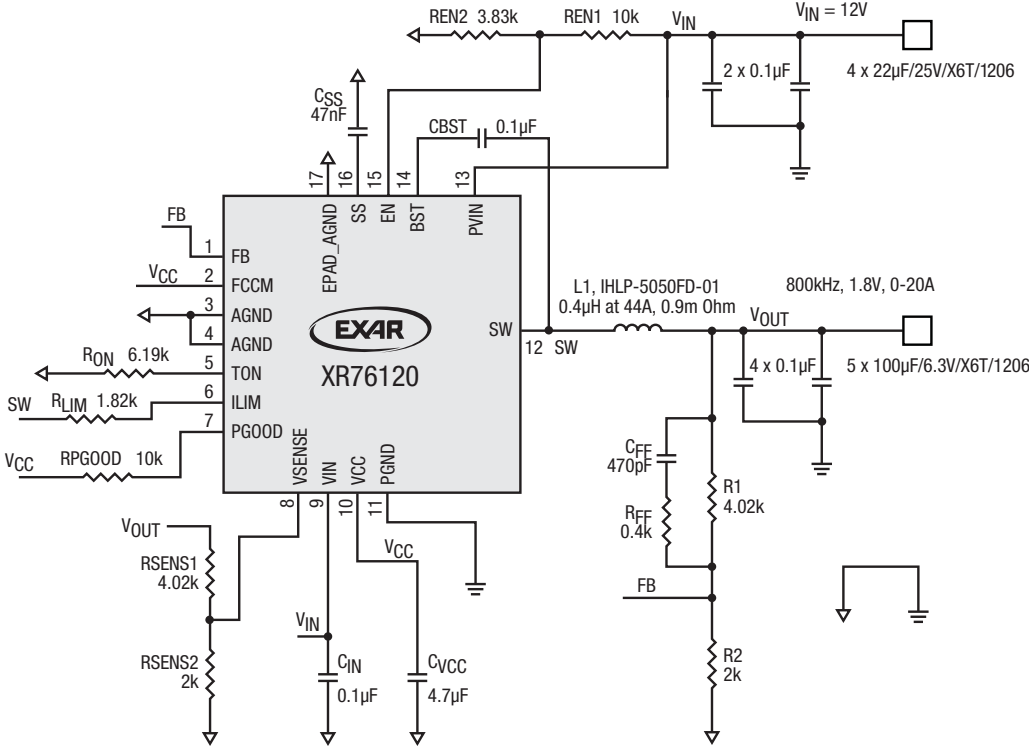


Figure 32. XR76120 Application Circuit Schematic

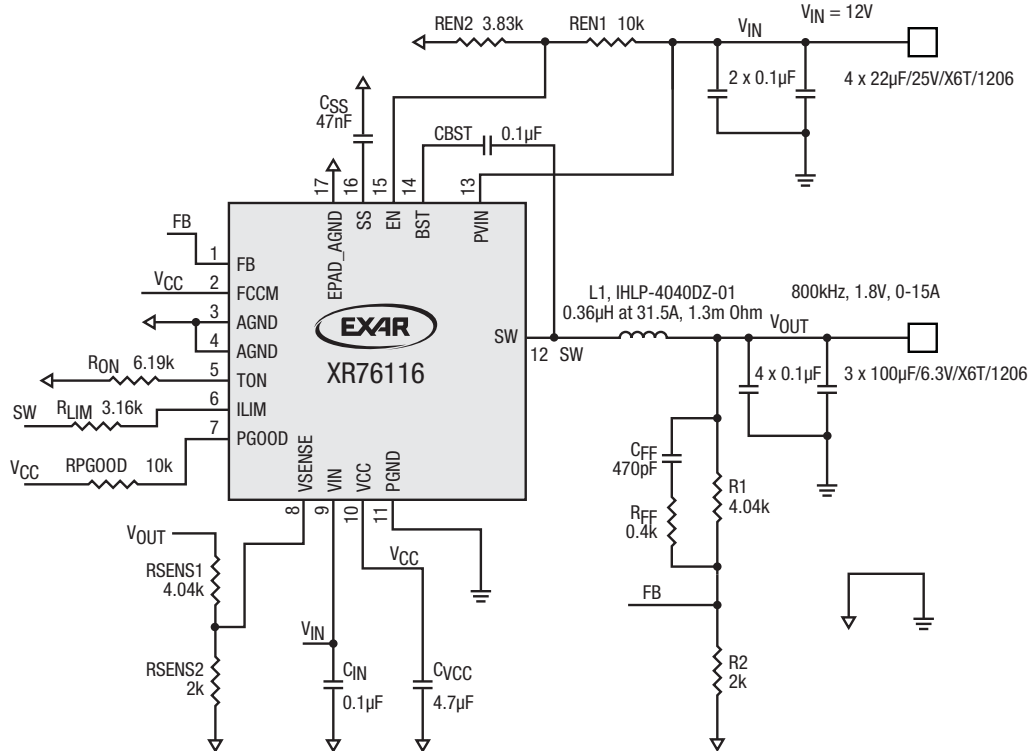
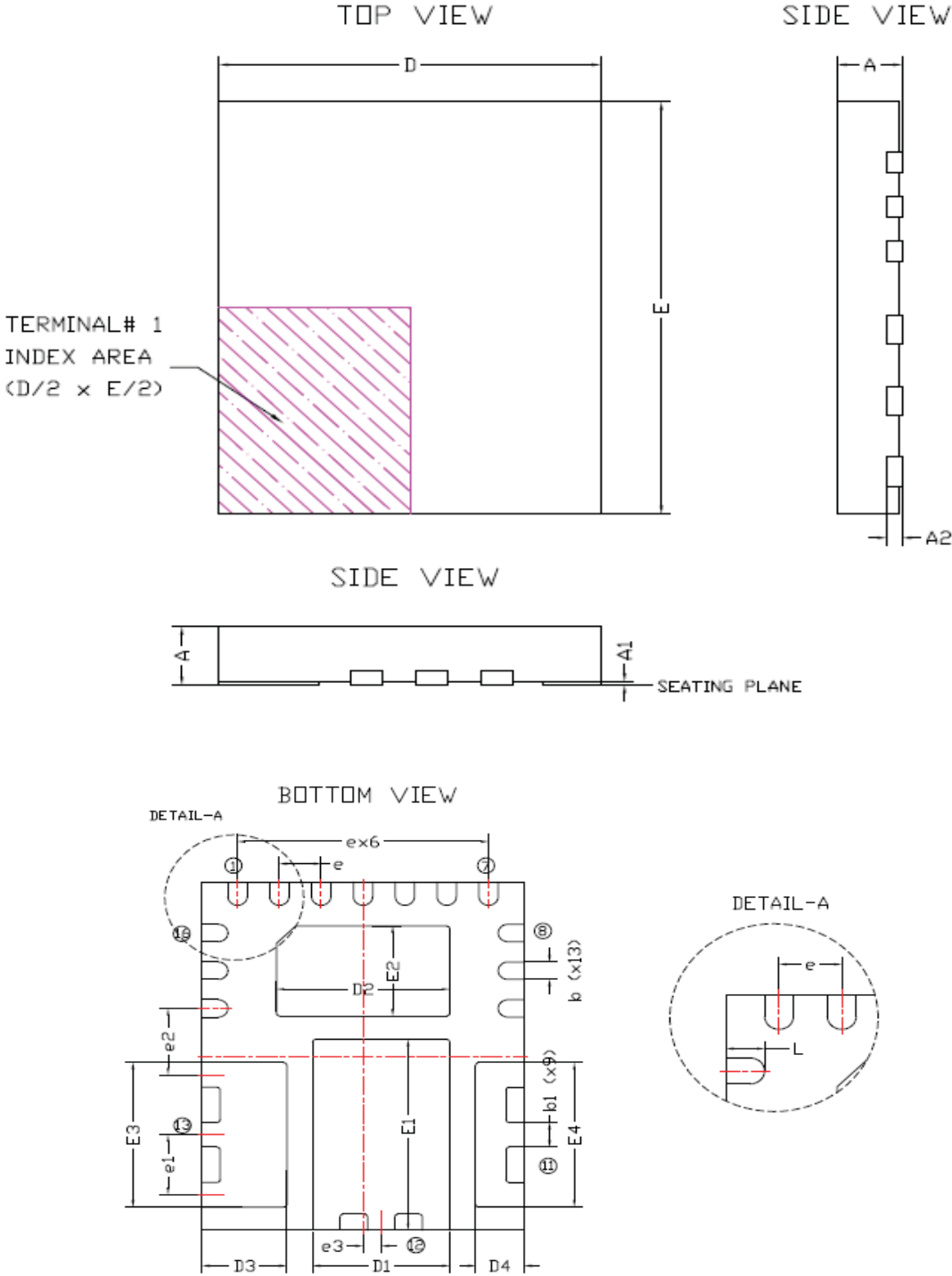


Figure 33. XR76116 Application Circuit Schematic

Package Description

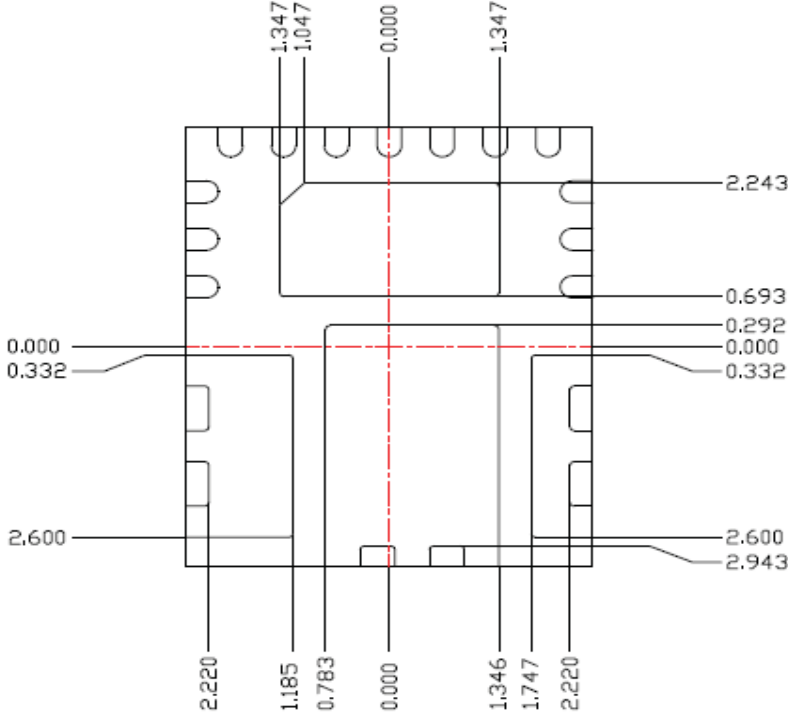


All dimensions are in mm and angles in degrees.

Figure 34. Package Description (1 of 3)

Package Description (Continued)

TERMINAL AND PAD EDGE DETAILS



| PACKAGE | 16L QFN 6x5 mm | | |
|---------|----------------|-------|-------|
| REF. | MIN. | NDM. | MAX. |
| A | 0.800 | 0.850 | 0.900 |
| A1 | 0 | - | 0,05 |
| A2 | 0,20 Ref. | | |
| b | 0.250 | 0.300 | 0.350 |
| b1 | 0.375 | 0.425 | 0.475 |
| D | 4.900 | 5.000 | 5.100 |
| D1 | 2.079 | 2.129 | 2.179 |
| D2 | 2.644 | 2.694 | 2.744 |
| D3 | 1.265 | 1.315 | 1.365 |
| D4 | 0.703 | 0.753 | 0.803 |
| E | 5.900 | 6.000 | 6.100 |
| E1 | 3.242 | 3.292 | 3.342 |
| E2 | 1.500 | 1.550 | 1.600 |
| E3 | 2.441 | 2.491 | 2.541 |
| E4 | 2.441 | 2.491 | 2.541 |
| e | 0,650 Ref. | | |
| e1 | 1,033 Ref. | | |
| e2 | 1,145 Ref. | | |
| e3 | 0,281 Ref. | | |
| L | 0.350 | 0.400 | 0.450 |

All dimensions are in mm and angles in degrees.

Figure 34. Package Description (2 of 3)

Order Information

| Part Number | Operating Temperature Range | Environmental Rating | Package | Packaging Quantity | Marking |
|----------------|--------------------------------|--------------------------------|---------------|--------------------|--------------------------------|
| XR76116EL-F | -40°C ≤ T _J ≤ 125°C | RoHS-compliant halogen free | 5mm x 6mm QFN | Bulk | XR76116EL YYWWF XXXXXXXX |
| XR76116ELMTR-F | | | | 250/tape and reel | |
| XR76116ELTR-F | | | | 3K/tape and reel | |
| XR76120EL-F | | | | Bulk | XR76120EL YYWWF XXXXXXXX |
| XR76120ELMTR-F | | | | 250/tape and reel | |
| XR76120ELTR-F | | | | 3K/tape and reel | |
| XR76116EVB | XR76116 evaluation board | | | | |
| XR76120EVB | XR76120 evaluation board | | | | |

NOTE:
YY = Year, WW = Work Week, F = Halogen Free, XXXXXXXX = Lot Number.



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